

# **APPENDIX**

## APPENDIX - I

CALCULATION OF Q

The experimental value of quality factor Q can be obtained by using the observed values of gain for low pass, high pass and band pass response<sup>1</sup>.

The second order transfer function for low pass, band pass and high pass filter are given by

$$H_{LP}(S) = \frac{G_{LP} \cdot w_0^2}{S^2 + (w_0/Q)S + w_0^2} \quad \dots \dots \dots (1.A)$$

$$H_{BP}(S) = \frac{G_{BP} (w_0/Q) \cdot S}{S^2 + (w_0/Q)S + w_0^2} \quad \dots \dots \dots (1.B)$$

$$H_{HP}(S) = \frac{G_{HP} \cdot S^2}{S^2 + (w_0/Q)S + w_0^2} \quad \dots \dots \dots (1.C)$$

Analysis of a new active -R biquadratic filter circuit given in section (3.3) of fig.(3.1) yields.

$$H_{LP}(S) = \frac{-(1/R_3)GB_1GB_2}{S^2[1/R_A + 1/R_2 + 1/R_3] + S[GB_1(1/R_2 + 2/(4R + R_1)) - GB_2(2/R_1 - 1/R_A)] + GB_1GB_2(2/R_1 - 1/R_A)} \quad \dots \dots \dots (2.A)$$

$$H_{HP}(s) = \frac{(1/R_3)s^2}{s^2[1/R_A + 1/R_2 + 1/R_3] + s[GB_1(1/R_2 + 2/(4R+R_1)) - GB_2(2/R_1 - 1/R_A)] + GB_1GB_2(2/R_1 - 1/R_A)} \dots \dots \dots (2.B)$$

$$H_{BP}(s) = \frac{-(1/R_3)s}{s^2[1/R_A + 1/R_2 + 1/R_3] + s[GB_1(1/R_2 + 2/(4R+R_1)) - GB_2(2/R_1 - 1/R_A)] + GB_1GB_2(2/R_1 - 1/R_A)} \dots \dots \dots (2.C)$$

Comparing numerator equancy (1) & (2), we have

$$G_{LP} \cdot W_o^2 = (1/R_3) \cdot GB_1 GB_2 \dots \dots \dots (3.A)$$

$$G_{HP} = (1/R_3) \dots \dots \dots (3.B)$$

$$G_{BP} (W_o/Q) = (1/R_3) \cdot GB_1 \dots \dots \dots (3.C)$$

By comparing denominator of equancy (2) with equancy (1)

$$(W_o/Q) = GB_1 [1/R_2 + 2/(4R+R_1)] - GB_2 (2/R_1 - 1/R_A) \dots \dots \dots (4)$$

$$W_o^2 = GB_1 \cdot GB_2 (2/R_2 - 1/R_A) \dots \dots \dots (5)$$

$$1 = 1/R_A + 1/R_2 + 1/R_3 \dots \dots \dots (6)$$

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where  $\frac{1}{R_A} = \frac{2}{R_1} - \frac{4R}{4R R_1 + R_1^2}$

Putting values of (4), (5), (6) in (3.A), (3.B), (3.C)

$$G_{BP} = \frac{1}{R_3 [1/R_2 + 2/(4R+R_1) - 4R/(4RR_1^2+R_1)]} \dots\dots\dots(7)$$

$$G_{LP} = \frac{4RR_1 + R_1^2}{4RR_3} \dots\dots\dots(8)$$

$$G_{HP} = (1/R_3) \dots\dots\dots(9)$$

From equation (5)

$$W_o^2 = G_B_1 \cdot G_B_2 \left( \frac{4R}{4RR_1 + R_1^2} \right)$$

Putting the value of  $W_o$  from equation (4) in this equation

we get

$$Q_{TH}^2 = \left( \frac{4R}{4RR_1 + R_1} \right) \left[ \frac{1}{\left( \frac{1}{R_2} + \frac{2}{4RR_1} - \frac{4R}{4RR_1 + R_1^2} \right)^2} \right] \dots\dots\dots(10)$$

Putting values from equation (7),(8),(9) in equation (10)

$$Q_{\text{EXP.}}^2 = \frac{G_{\text{BP}}^2}{G_{\text{HP}} \cdot G_{\text{LP}}} \dots\dots\dots(11)$$

Where TH for theoretical and EXP. for experimental. This equation is used to calculate the value of Q for the bandpass response ;where  $R_1$  is at center tapped.

For the variation of the center tap ,

$$\begin{aligned} Q_{\text{DE}}^2 &= \\ &= \frac{R}{R_1 [(1-A)AR_1 + R]} \cdot \frac{1}{[\frac{1}{R_2} + \frac{A}{(1-A)AR_1 + R} - \frac{R}{R_1 [(1-A)AR_1 + R]}]^2} \\ &\dots\dots\dots(12) \end{aligned}$$

$$Q_{\text{EXP.}}^2 = \frac{G_{\text{BP}}^2}{G_{\text{LP}} \cdot G_{\text{HP}}} \dots\dots\dots(13)$$

REFERENCE

1. MOHAN N. and PATIL R.L. " An analytical method for determining the Q - Values of an state - variable active-R filter", Journal of instrumentation society of India, VOL. - 176, (1987) 306 - 312.

10 :REM  
20 :REM  
30 :REM  
40 :REM  
50 :REM  
60 :REM  
70 :REM  
80 :REM  
90 :REM

## APPENDIX II

100 INPUT "INPUT THE DESIGN VALUE OF FD ";FD  
200 INPUT " INPUT THE VALUE OF R ";R  
300 INPUT " INPUT THE DESIGN VALUE OF Q ";Q  
400 PRINT  
500 PRINT TAB(22); "D=";D,: PRINT TAB(46); "Fd=";FD,  
600 GB=7.9\*44\*1000000! /7  
700 F1=(GB\*GB\*49+4\*R/(44\*44\*FD\*FD))  
800 R1=(-2\*R+SQR(4\*R\*R+F1))  
900 RA=R1/2+(R1+R)/(R1+2\*R)  
1000 I1=R1\*(R1+4\*R) : I2=2\*(R1-2\*R)  
1100 R2= 1/((44\*FD)/(GB\*7\*Q)-I2/I1)  
1200 R3= 1/((44\*FD)/(GB\*7\*Q)-R1/(RA\*(R1+2\*R)))  
1300 R4=1/((1-1/(RA))-1/(R2))  
1400 R=R4\*100  
1500 R1=INT(100\*R1) : R2=INT(100\*R2) : R3=INT(100\*R3) : RA=INT(100\*RA)  
1600 PRINT TAB(65); "R=";R  
1700 PRINT  
1800 PRINT TAB(22); "R1=";R1,  
1900 PRINT TAB(46); "R2=";R2,  
2000 PRINT TAB(65); "R3=";R3  
2100 PRINT  
2200 PRINT TAB(30); "FREQUENCY"  
2300 PRINT TAB(22); "LOW PASS"  
2400 PRINT TAB(46); "HIGH PASS"  
2500 PRINT TAB(65); " BAND PASS"  
2600 PRINT  
2700 PRINT TAB(1); F,  
2800 D1=(44\*FD/7)^2 : D2=(44\*F/7)^2  
2900 D=SQR((D1-D2)^2+(D1\*D2)/(Q\*Q))  
3000 TPRINT  
3100 T1=(GB\*88)/(R3\*D) : T1=((20\*LOG(T1))/2.303) : PRINT TAB(24); T1,  
3200 T2=D2/(R3\*D) : T2=((20\*LOG(T2))/2.303) : PRINT TAB(46); T2,  
3300 T3=(44\*F\*GB)/(7\*D\*R3) : T3=-((20\*LOG(T3))/2.303) : PRINT TAB(65); T3,  
3400 NEXT N  
3500 L=104L  
3600 NEXT N  
3700 END

## APPENDIX - 3

TABLE - 1

Resistance values of new active -R filter  
circuits for different value of Q

$F_0$ kHz	A	Q	Designed value in $\Omega$			Experimental Value in $\Omega$		
			$R_1$	$R_2$	$R_3$	$R_1$	$R_2$	$R_3$
10	0.5	0.2	20 k	1.8k	106	20k	1.6k	99.3
10	0.5	0.5	20 k	6.2k	102.	20k	6.2k	99.3
10	0.5	1.0	20 k	30 k	101	20k	29k	99.3

TABLE-2

Resistance values of new active -R filter circuits  
with variation of center frequency  $F_0$

$F_0$ kHz	A	Q	Desined value in $\Omega$			Experimental value in $\Omega$		
			$R_1$	$R_2$	$R_3$	$R_1$	$R_2$	$R_3$
10	0.5	1	20k	30.1k	101.3	20.1k	29.6k	99.3
50	0.5	1	3.7k	4.35k	107.7	3.59k	4.29k	109
100	0.5	1	1.7k	1.62k	119.2	1.67k	1.61k	118

TABLE-3

Resistance value of new active -R filter circuits for  
variation of tapping point.

A	$F_0$ KHz	Q	Designed values in $\Omega$				Expt.values in $\Omega$			
			$AR_1$	$(1-A)$ . $R_1$	$R_2$	$R_3$	$AR_1$	$(1-A)$ . $R_1$	$R_2$	$R_3$
0.1	10	0.2	2K	18K	1.8K	106.8	2.18K	17.7K	1.61K	99.3
0.3	10	0.2	6K	14K	1.8K	106.8	6.2K	14.1K	1.61K	99.3
0.5	10	0.2	10K	10K	1.8K	106.8	9.91K	10.2K	1.61K	99.3
0.7	10	0.2	14K	6K	1.8K	106.8	14.1K	6.2K	1.61K	99.3
0.9	10	0.2	18K	2K	1.8K	106.8	17.7K	2.1K	1.61K	99.3